

Error Rate Analysis for Coded Multicarrier Systems over Quasi-static Fading Channels

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- 1 Overview of Concepts
- 2 Performance Analysis of Coded OFDM Systems
 - Error Vectors
 - Pairwise Error Probability
 - Per-realization BER
 - Direct Average BER
- 3 Performance Results for Multiband OFDM

Orthogonal Frequency Division Multiplexing (OFDM)

- Many orthogonal subchannels via IFFT & FFT
- (Punctured) convolutional codes for error protection
- Robust in frequency-selective fading

Quasi-Static Fading Channel

- Fading conditions constant for “relatively long” period of time
- Here: “relatively long” = (at least) one packet
- **Can only code over one channel realization**
- Channel is also frequency-selective

Analysis: Performance Measures

We will consider two different performance measures

Average Bit Error Rate (BER)

Average BER of a (large) number of different channel realizations



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Outage BER

- if we assume $X\%$ of channels are bad (“in outage”), what is the worst-case performance of the remaining $(100 - X)\%$ of the channels?
- (we will consider $X = 10\%$ outage probability)

Outage BER is a more relevant measure for quasi-static channels...
BUT, it is hard to determine.



Analysis: BER for Quasi-static Channels

Average BER for Fast Fading Channel

- Get pairwise error probability $P_2(d|h)$ for given h
- Integrate over pdf of h to get $P_2(d)$, then apply union bound

$$\text{BER} \leq \frac{1}{k} \sum_{d=d_{\text{free}}}^{\infty} \beta_d P_2(d)$$

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Quasi-static Channel

Each coded block transmitted over one channel realization

- Only “sees” a limited number of channel gains
- (and, correlation between bits depends on starting position)

CANNOT simply integrate over pdf of fading distribution!

Question

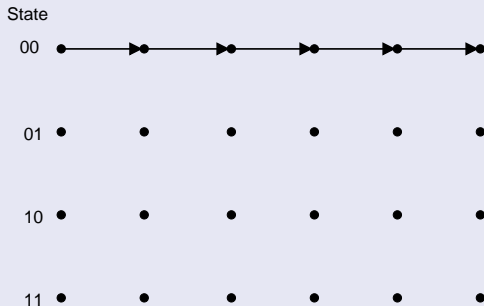
Can we predict the BER of a given realization?



Analysis: Error Vectors

Trellis representation of $R_c = 1/2$ (7,5) convolutional code

An example error path



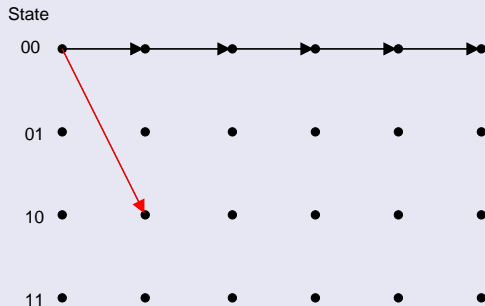
Input: { , , , }

Output \mathbf{e}_1 : { , , , , , , , }

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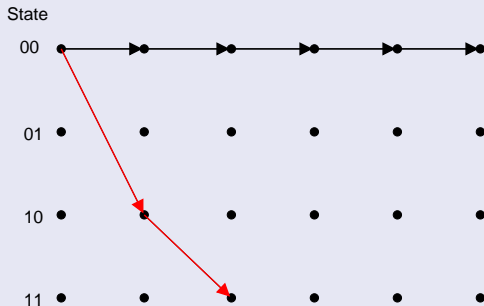
Input: {1, , , }

Output e_1 : {1, 1, , , , , }

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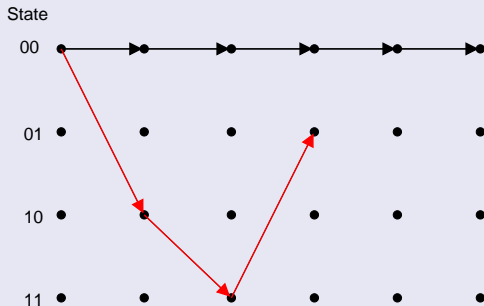
Input: $\{1, 1, \ , \}$

Output \mathbf{e}_1 : $\{1, 1, 0, 1, \ , \ , \}$

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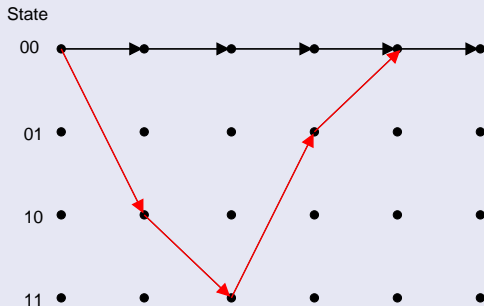
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Output e_1 : $\{1, 1, 0, 1, 0, 1, 1, 1\}$

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- Call this set $\mathcal{E} = \{\mathbf{e}_1, \dots, \mathbf{e}_L\}$
- l_ℓ is the length of \mathbf{e}_ℓ
- a_ℓ is the number of information bit errors associated with \mathbf{e}_ℓ

\mathcal{E} contains all the “most likely” error events (the ones with shortest length)

Frequency-domain System Model

$$\mathbf{r} = \mathbf{H}\mathbf{x} + \mathbf{n}$$

where

- \mathbf{x} are the transmitted symbols
- $\mathbf{H} = \text{diag}(\mathbf{h})$ is a (diagonal) matrix of channel gains
- \mathbf{n} are AWGN noise variables

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- For each error vector \mathbf{e}_ℓ , we form the “full” **error codeword** (length L_c)

$$\mathbf{q}_{i,\ell} = \underbrace{[0 \ 0 \ \dots \ 0]}_{i-1} \underbrace{[\mathbf{e}_\ell]}_{l_\ell} \underbrace{[0 \ 0 \ \dots \ 0]}_{L_c - l_\ell - i + 1}^T$$

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- Now, given $\mathbf{q}_{i,\ell}$, the **competing codeword** is given by

$$\mathbf{v}_{i,\ell} = \mathbf{c} \oplus \mathbf{q}_{i,\ell}$$



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- true codeword $\mathbf{c} \longrightarrow \mathbf{x}$
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- the PEP for the ℓ th error vector starting in the i th position, i.e., the probability that $\mathbf{v}_{i,\ell}$ is detected given that \mathbf{c} was transmitted, is given by

$$\text{PEP}_{i,\ell}(\mathbf{H}) = \Pr \{ \|\mathbf{r} - \mathbf{H}\mathbf{x}\|^2 > \|\mathbf{r} - \mathbf{H}\mathbf{z}_{i,\ell}\|^2 \mid \mathbf{H} \}$$

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- For given \mathbf{H}, i, ℓ

PEP

$$\text{PEP}_{i,\ell}(\mathbf{H}) = Q \left(\sqrt{\frac{\|\mathbf{H}(\mathbf{x} - \mathbf{z}_{i,\ell})\|^2}{2\mathcal{N}_0}} \right)$$

Analysis: Per-realization BER

BER for \mathbf{e}_ℓ , starting in the i th position, is given by

$$P_{i,\ell}(\mathbf{H}) = a_\ell \cdot \text{PEP}_{i,\ell}(\mathbf{H})$$

Sum over all $\mathbf{e}_\ell \rightarrow$ BER for the i th position

$$P_i(\mathbf{H}) = \sum_{\ell=1}^L P_{i,\ell}(\mathbf{H})$$

All positions equally likely \rightarrow BER $P(\mathbf{H})$ is

$$P(\mathbf{H}) = \frac{1}{L_c} \sum_{i=1}^{L_c} \min \left[\frac{1}{2}, \sum_{\ell=1}^L P_{i,\ell}(\mathbf{H}) \right]$$

This is the (approximate) BER for a given \mathbf{H}

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Once we know $P(\mathbf{H})$, we can easily obtain both average and outage BER



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Average BER

$$P_{\text{avg}} = \frac{1}{N_c} \sum_{j=1}^{N_c} P(\mathbf{H}_j)$$

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Average BER

$$P_{\text{avg}} = \frac{1}{N_c} \sum_{j=1}^{N_c} P(\mathbf{H}_j)$$

Outage BER

$$P_{\text{out}} = \max_{\mathbf{H}_j \in \mathcal{H}_{\text{in}}} P(\mathbf{H}_j)$$

(where \mathcal{H}_{in} are the channels that are NOT in outage)

Analysis: Direct Average BER

A Special Case

If channel gains \mathbf{h} are

- Rayleigh-distributed r.v.s, and
- Correlated with correlation matrix $\Sigma_{\mathbf{h}\mathbf{h}}$

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We only need to consider non-zero entries of $(\mathbf{x} - \mathbf{z}_{i,\ell})$ \rightarrow make

- \mathbf{x}' , $\mathbf{z}'_{i,\ell}$, $\mathbf{H}' = \text{diag}(\mathbf{h}')$, and \mathbf{n}'
- $\Sigma_{\mathbf{h}'\mathbf{h}'}$
- $\mathbf{D} = \text{diag}(\mathbf{x}' - \mathbf{z}'_{i,\ell})$
- $\mathbf{g} = \mathbf{H}'(\mathbf{x}' - \mathbf{z}'_{i,\ell}) = \mathbf{D}\mathbf{h}'$

Analysis: Direct Average BER

$$\begin{aligned}\mathbb{E}(\mathbf{g}) &= \mathbf{0}_{\eta \times 1} \\ \mathbb{E}(\mathbf{g}\mathbf{g}^H) &= \mathbf{R}_{\mathbf{g}\mathbf{g}} = \mathbf{D}\boldsymbol{\Sigma}_{\mathbf{h}'\mathbf{h}'}\mathbf{D}^H\end{aligned}$$

i.e., the distribution of \mathbf{g} is zero-mean complex Gaussian with covariance matrix $\mathbf{R}_{\mathbf{g}\mathbf{g}}$.

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i.e., the distribution of \mathbf{g} is zero-mean complex Gaussian with covariance matrix $\mathbf{R}_{\mathbf{g}\mathbf{g}}$. Following [Veervalli '01], we can write the bit error probability for the j^{th} error vector starting in the i^{th} position as

$$\bar{P}_{i,\ell} = \frac{a_\ell}{\pi} \int_0^{\pi/2} \left[\det \left(\frac{E_s \mathbf{R}_{\mathbf{g}\mathbf{g}}}{\mathcal{N}_0 \sin^2 \theta} + \mathbf{I} \right) \right]^{-1} d\theta$$

Sum over ℓ and average over i as before $\rightarrow \overline{\text{BER}}$ is

$$\bar{P} = \frac{1}{L_c} \sum_{i=1}^{L_c} \bar{P}_i = \frac{1}{L_c} \sum_{i=1}^{L_c} \sum_{\ell=1}^L \bar{P}_{i,\ell}$$

This is the (approximate) average BER

Results: MB-OFDM System and Channel Models

We will consider Multiband OFDM as an example OFDM system

MB-OFDM

- Multiband OFDM is a leading high rate UWB system
- 3.1–10.6 GHz \rightarrow 14 subbands of 528 MHz
- QPSK modulation (also consider 16-QAM)
- Punctured convolutional code

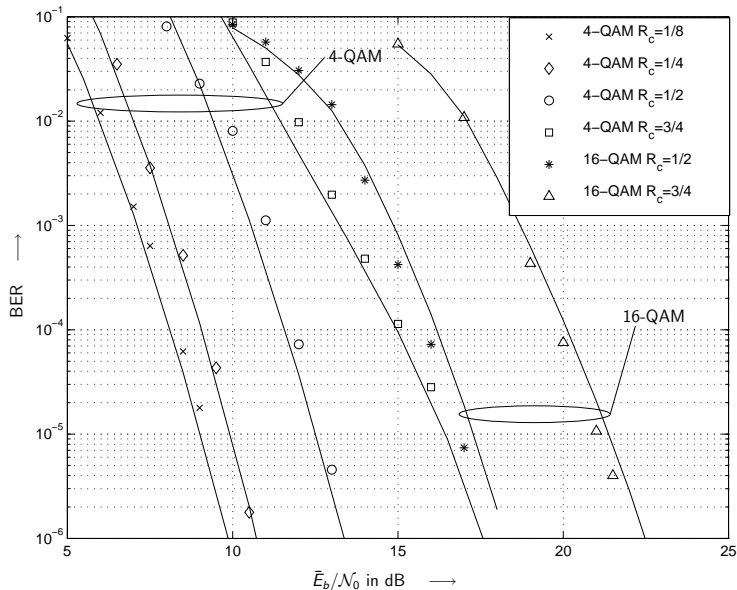
Channel impulse response (802.15.3a model)

$$h(t) = X \sum_{l \geq 0} \sum_{k \geq 0} \alpha_{k,l} \delta(t - T_l - \tau_{k,l})$$

- Impulse response consists of clusters of multipath components
- We will consider equivalent frequency-domain channel \mathbf{H}
- Channel is quasi-static
- For given X , gains \mathbf{H} are complex Gaussian

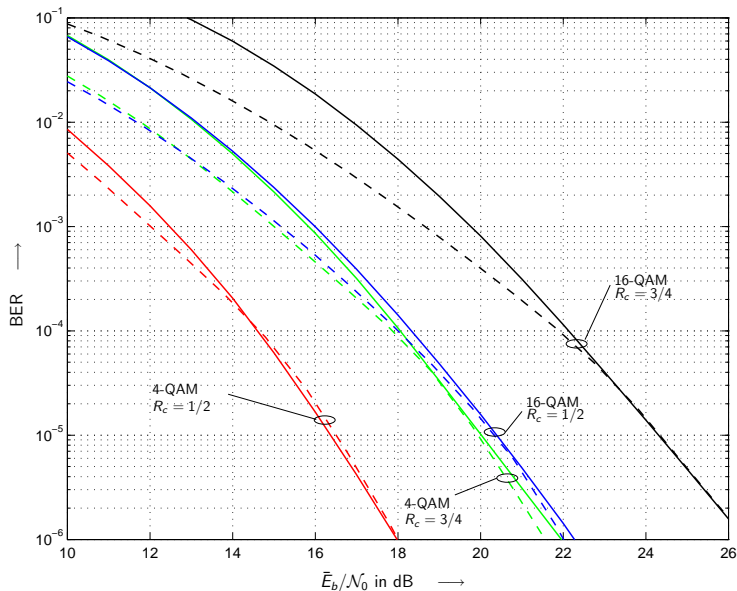


Results: Outage BER



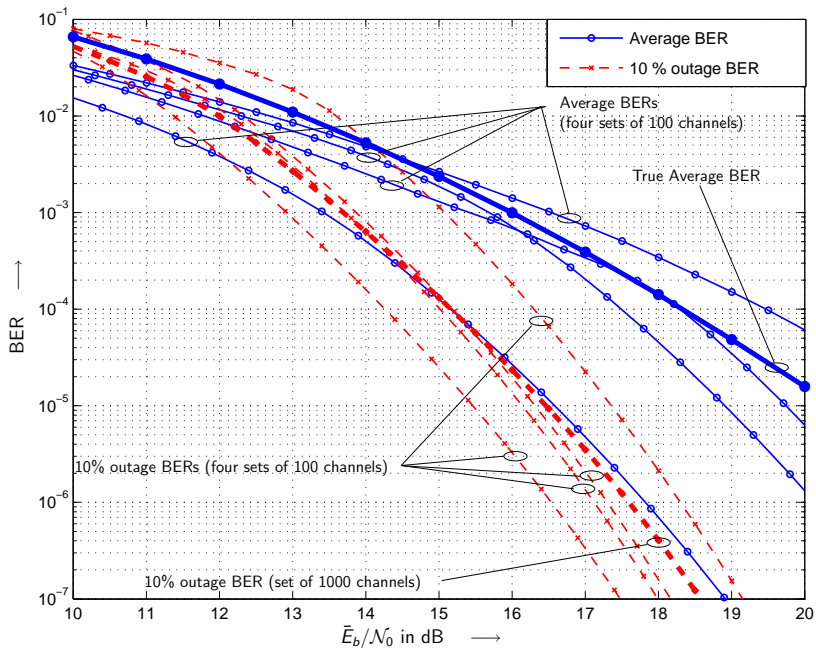
10% Outage BER. 100 CM1 channel realizations.

Results: Average BER



Solid: Direct average. Dashed: average over 10,000 channel realizations.

Results: Different Channel Realizations



Conclusions

Analysis

Developed methods to estimate

- Outage BER
- Average BER

These methods apply to any OFDM system in quasi-static channel

Results

- Can approximate OFDM performance without simulations
- Both average and outage BER easily available
- Small set of channel realizations \longrightarrow misleading results

More detailed journal preprint at www.ece.ubc.ca/~csnow/

